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## COMPUTER ANALYSIS OF MAGNETIC DEVICES OF INFORMATION CONTROL SYSTEMS AUTOMATICS

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### ABSTRACT

In the paper the possibility of computer analysis of quite complex magnetic devices used in information control systems by one of the numerical methods, by the method of secondary sources, is considered. The proposed calculation algorithm allows the use a variety of computing systems and this can significantly reduce the complexity of the process and allow virtually less error. In this case, it is possible to effectively apply this method for wider range of various magnetic elements and devices of information control systems automatics.

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## INTRODUCTION

Existing various mathematical methods for calculating applied technical problems stipulates wideuse of computer technology, which promotealgorithmization of these calculations, significant reduction in labor-intensiveness and a decrease in error. From this point of view, in order to calculate quite complex problems as magnetic elements and devices of automatics of control systems, various numerical methods meet these goals. One of these numerical methods is the method of secondary sources [1-7].

This method allows to completely algorithmize the calculation process based on the application of computing technology and at the same time to determine of magnetic elements and devices the flow distribution, electric and magnetic parameters of these elements on each  $i$ - th elementary area. In this connection we consider calculation of magnetic elements and devices by the method of secondary sources (MSS)

**Calculation of magnetic elements and devices by the method of secondary sources:** According to the method of secondary sources, calculation of electromagnetic system is considered in the form of the system of integral equations by which secondary sources of the field are determined in the form: a) density of fictitious magnetic charges ( $\sigma_M : \delta$ ), density of eddy currents  $\delta v$ . Having determined secondary sources, distribution of electromagnetic field is determined, and knowing electromagnetic fielddistribution at each point of the considered magnetic system, integral characteristics of the given devices are determined by the known relations. According to (195), calculated integral equations have the form:

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$$\sigma_M(Q) + \alpha_m \oint_{S_\Phi} \sigma_M(M) \left( \frac{r_{Q_m n Q}}{2\pi r_{Q_m}^3} - \frac{1}{S_\Phi} \oint_{S_\Phi} \frac{r_{QH n Q}}{2\pi r_{Q_m}^3} \right) = d\overline{S}_m = \dots\dots\dots(1.1)$$

$$= -\mu_0 \lambda_M \int_{V_K} \overline{\delta}(M) \frac{r_{Q_m n Q}}{2\pi r_{Q_m}^3} dV_\mu - \mu_0 \lambda_M \int_{V_K} \overline{\delta}_b(M) \frac{r_{Q_m n Q}}{2\pi r_{Q_m}^3} dV_\mu$$

$$\overline{B}(Q) = \frac{\mu_0}{4\pi} \int_{V_K} \frac{\overline{\delta}(M) r_{Q_m}}{r_{Q_m}^3} dV_M - \frac{\mu_0}{4\pi} \int_{V_K} \frac{[\overline{\delta}_b(M) r_{Q_m}]}{r_{Q_m}^3} dV_M -$$

$$- \frac{1}{4\pi} \oint_S \frac{\sigma_H(M) r_{Q_m}}{r_{Q_m}^3} dS_M \dots\dots\dots(1.2)$$

$$\sigma(Q) + \oint_S \sigma(M) \left( \frac{r_{Q_m n a}}{2\pi r_{aM}^3} - \frac{1}{S} \oint_S \frac{r_{Q_m n Q}}{r_{Q_m}^3} d\overline{S}_M \right) = \frac{j\omega}{2\pi} \int_V \overline{B}(M) \frac{[r_{Q_m n Q}]}{r_{Q_m}^3} dV_m \dots\dots\dots(1.3)$$

$$\overline{\delta}_b(Q) = \frac{j\omega\gamma'\gamma_0}{4\pi} \int_V \frac{[r_{Q_m} \overline{B}(M)]}{r_{Q_m}^3} dV_m - \frac{\gamma'\gamma_0}{4\pi} \oint_S \sigma(M) \frac{r_{Q_m}}{r_{Q_m}^3} dS_m \dots\dots\dots(1.4)$$

where  $r_{Q_M} = (X_M - X_Q)\vec{i} + (Y_M - Y_Q)\vec{j} + (Z_M - Z_Q)\vec{k}$  is the vector of distance between the point  $Q$ , where the values of the desired value are determined, and the point  $M$ , where the source is located;

$\sigma_M$  is the surface density of the fictitious magnetic charge (source of the field of magnetization of magnetic charges);

$$\lambda_M = \frac{\mu'_i - \mu'_j}{\mu'_i + \mu'_j} \text{ is constant;}$$

$\mu'_i, \mu'_j$  are relative magnetic permeabilities at the beginning and at the end of the area;

$S_Q$  is the surface of the magnetic core;

$V_K$  is the volume of magnetizing coils;

$\overline{\delta}$  is the current density in the magnetization core (primary source);

$\overline{\delta}_b$  is the density of eddy currents (secondary source);

$\sigma$  is the source of harmonic charge function;

$\gamma'$  is the relative conductivity of medium;

$\gamma_0$  is the conductivity of homogeneous unbounded medium.

The considered integral equations (1.1-1.4) determine mathematical model of electromagnetic field in piecewise-homogeneous conducting and ferromagnetic field. According to [1], the algorithm for solving the written equations is represented in the following way:

1. Assuming  $\overline{\delta}_b = 0, \overline{\sigma}_M = 0$ , determine  $\overline{B}_0$  according to (1.2).
2. Knowing  $\overline{B}_0$  by the magnetization curve of the material  $B = f(H), \overline{H}_0$ , we determine and then  $\mu'_0 = \frac{B_0}{H_0}$ .
3. Solve equation (1.3) for  $\overline{B} = \overline{B}_0$ , having determined the distribution  $\sigma^{(1)}$ .
4. Solve equation (1.4) i.e. determine  $\overline{\delta}_b^{(1)}$ .

5. Solve equation (1.1) i.e. having determined  $\sigma_M^{(1)}$ .

The given steps of the algorithm are repeated until the iterative process converges i.e.  $\vec{B} = \vec{B}_0 + \vec{B}^{(1)} + \dots$

With the convergence of the iterative process, the magnetic flux is determined at each  $i$ -th area according to the relation:

$$\Phi_i = \frac{1}{4\pi} \oint_L \sigma(M) \frac{|\vec{r}_{Q_m V}|}{r_{Q_M} (r_{Q_M} - r_{Q_m V})} d\vec{S}_m d\vec{l}_Q \dots\dots\dots(1.5)$$

The solution of the written integral equations (1.1.-1.4) is realized by the method of successive approximations. The written expressions are reduced to the solution of a linear integral equation of the form

$$\sigma(Q) + \frac{\lambda}{2\pi} \oint_S K(Q, M) \sigma(M) dS_m = f(Q) \dots\dots\dots(1.6)$$

where  $K(Q, M) = \frac{r_{Q_m n_Q}}{r_{Q_m}^3} - \frac{1}{S} \oint_S \frac{r_{Q_m n_Q}}{r_{Q_m}^3} dS_m$  is the equation core;

- $\sigma(Q)$  is an unknown (desired) function,
- $f(Q)$  is a free member,
- $j$  is a numerical parameter.

If we divide the entire closed surface of integration  $S$  into  $n$  rather small areas of square  $\Delta S_j$  centered at the points,  $M_j (j = 1, 2, \dots, n)$ , then the integral contained in expression (1.6), may be represented in the form of the finite sum of integrals in areas  $\Delta S_j$ .

$$\oint_S K(Q, M) \sigma(M) dS_m = \sum_{j=1}^n \int_{\Delta S_j} K(Q, M) \sigma(M) dS_m \dots\dots\dots(1.7)$$

Assuming that on elementary areas the function  $\sigma(M)$  is constant and equals the value  $\sigma(M_j)$  at the point  $M_j$ , we obtain the following approximate expression for the integral (1.6):

$$\oint_S K(Q, M) \sigma(M) dS_m \approx \sum_{j=1}^n G(Q, M_j) \sigma(M_j) \dots\dots\dots(1.8)$$

$$G(Q, M_j) = \int_{\Delta S_j} K(Q, M) \sigma(M) dS_m \dots\dots\dots(1.9)$$

Having substituted (1.8) in (1.6) we obtain the equality:

$$\sigma(Q) + \frac{\lambda}{2\pi} \sum_{j=1}^n G(Q, M_j) \sigma(M_j) = f(Q) \dots\dots\dots(1.10)$$

Expression (1.10) for the points  $Q_i$  the centers of the areas  $\Delta S_i, (i = 1, 2, \dots, n)$  is written in the form of the system of linear algebraic equations:

$$\sigma(Q_i) + \frac{\lambda}{2\pi} \sum_{j=1}^n G(Q_i, M_j) \sigma(M_j) = f(Q_i) \quad i = 1, 2, \dots, n \dots\dots\dots(1.11)$$

$$G(Q_i, M_j) = \int_{\Delta S_j} \left( \frac{r_{Q_m n_Q}}{r_{Q_m}^3} - \frac{1}{S} \oint_S \frac{r_{Q_m n_Q}}{r_{Q_m}^3} dS_m \right) dS_m, \dots\dots\dots(1.12)$$

that satisfy the values of the desired  $\sigma(Q_i)$  function at the points  $dS_M$ .

The  $n$ -th order system (1.12) is written in the form:

$$\sum_{j=1}^n a_{ij} \sigma_j = f_i \quad i = 1, 2, \dots, n \dots\dots\dots(1.13)$$

or in the vector matrix form

$$\overline{A}\overline{\sigma} = \overline{f} \tag{1.14}$$

where  $\sigma_j = \sigma(M_j)$  are the components of  $n$ -dimensional vector  $\overline{\sigma}$ ,  $f_i = f(Q_i)$  are the components of  $n$ -dimensional vector  $\overline{f}$ ,  $a_{ij}$  are the elements of the matrix  $A$ .

The  $n$ -th order expression (1.14) is solved by the method of block iterations that allows more economical use of computer-on-line storage.

The matrix  $A$  is divided into  $p^2$  rectangular cells  $\overline{A}_{ij}$  ( $i = 2, 3, \dots, p$ ). Dimension of rectangular matrices  $\overline{A}_{ij}$  equals  $(m_i \times m_j)$ , where  $\sum_{i=1}^p m_i = n$ . The  $n$ -dimensional vectors  $\overline{\sigma}$  and  $\overline{f}$  are divided into  $p$  vectors  $\overline{\sigma}_i$  and  $\overline{f}_i$  of dimension  $m_i$ :

$$\overline{\sigma} = \{\overline{\sigma}_1, \overline{\sigma}_2, \dots, \overline{\sigma}_i, \dots, \overline{\sigma}_p\} \tag{1.15}$$

$$\overline{f} = \{\overline{f}_1, \overline{f}_2, \dots, \overline{f}_i, \dots, \overline{f}_p\} \tag{1.16}$$

Then the system (1.14) will be equivalent to the systems

$$\overline{A}_{ij}\overline{\sigma} = \overline{f}_i \quad i = 1, 2, \dots, p \tag{1.17}$$

each of order  $m_i$ .

The iterative process is constructed in such a way that at each step  $p$  systems are solved by the prime method and the order of each systems is less than  $n$ .

Having set the initial approximation of vectors  $\sigma_i^{(0)}$  ( $i = 2, 3, \dots, p$ ) we determine  $\overline{\sigma}_i^{(1)}$ , solving by the primary method the first one from the systems (1.17) of  $m_i$  - order:

$$\overline{\sigma}_i^{(1)} = \overline{A}_{11}^{-1} \left( \overline{f}_1 - \sum_{j=2}^p \overline{A}_{1j} \overline{\sigma}_j^{(0)} \right) \tag{1.18}$$

where  $\overline{A}_{11}^{-1}$  - is a matrix inverse to square matrix  $\overline{A}$ .

Then from the second system of (1.17) the  $m_2$  order first approximation  $\overline{\sigma}_2^{(1)}$ ,

$$\overline{\sigma}_2^{(1)} = \overline{A}_{22}^{-1} \left( \overline{f}_2 - \overline{A}_{21} \overline{\sigma}_1^{(1)} - \sum_{j=3}^p \overline{A}_{2j} \overline{\sigma}_j^{(0)} \right) \tag{1.19}$$

For the  $i$ -value of the first approximation of the vector  $\overline{\sigma}_i^{(1)}$ , we have:

$$\overline{\sigma}_i^{(1)} = \overline{A}_{ii}^{-1} \left( \overline{f}_i - \sum_{j=1}^{i-1} \overline{A}_{ij} \overline{\sigma}_j^{(1)} - \sum_{j=i+1}^p \overline{A}_{ij} \overline{\sigma}_j^{(0)} \right) \quad i = 1, 2, \dots, p \tag{1.20}$$

where  $\overline{A}_{ii}^{-1}$  is a matrix inverse to the matrix  $\overline{A}_{ii}$  - of order  $m_i$ .

The approximation of the  $k$ -th step of the iterative process is:

$$\overline{\sigma}_i^{(k)} = \overline{A}_{ii}^{-1} \left( \overline{f}_i - \sum_{j=1}^{i-1} \overline{A}_{ij} \overline{\sigma}_j^{(k)} - \sum_{j=i+1}^p \overline{A}_{ij} \overline{\sigma}_j^{(k-1)} \right) \quad i = 1, 2, \dots, p. \tag{1.21}$$

Information about geometry of magnetic core, magnetizing coils (sizes, width, thickness, length) the basic magnetization curve enter in the form of the approximating function  $B(H) = aH^b(1+cH)$  for electrical steel  $\text{ЭА-1}$  the values of constants  $a = 0,99483$ ,  $b = 0,123398$ ,  $c = 000092$ . The magnetization curve may be given in the form of a table  $B = f(H)$ .

The values of densities of currents of sources, square of the magnetic core surface are set up. Volume and surface of magnetic elements and devices, and also volume of magnetization coils is divided into  $N$  number of rectangular areas. The flow distribution value is determined in the centre of these areas and in a such a way we obtain the dependence  $\Phi_i = f(N)$ , where  $N$  is the number of the points (of the centers of the  $i$  - th rectangular areas of magnetic elements). The coordinates of these points are determined with respect to the given system of coordinates and enter into the memory of a computer. Implementation of the program of calculation of magnetic elements and devices is performed according to block – diagram of the calculation program (Fig. 1). Description of the given block-diagram of the calculation program is given in table 1.

## 2. Description of block – diagram of the calculation program of magnetic elements and devices by the method of secondary sources (MSS).

1. Initial data are set up;

$N$  is the number of elementary segments in the contour  $l$ ;

$X_1, Y_1$  are the coordinates of the location point of the first given charge (current),

$X_2, Y_2$  are the coordinates of the location point of the second given charge (current),

$MAG$  is a sign equal to 0 in the case of an electric field and to 1 in the case of magnetic field;

$KZ$  is the number of the versions of the problem and equals the number given by magnetic permeability of medium inside and outside the contour  $l$ .

2. The number of the points  $M$  of the contour  $l$  is calculated.

3. The arrays  $X, Y, D, A, B, C, CP, R2M, CIN$  are calculated.

4. The coordinates of extreme points of segments  $X_i, Y_i$  are entered.

5. Organization of the input cycle of coordinates  $X_i, Y_i$ , changing from 1 to  $i$  through 1.

6.  $LK = 0$  – perimeter of the contour  $l$  of approximated segment is calculated.

7. The distance between the points by  $X - Y A_i$  and  $Y - Y A_i$  the distance  $D_i$  between them are calculated.

8.  $LK = \sum D_i, A_i = \frac{A_i}{D_i}, B_i = \frac{B_i}{D_i}, X_i = (X_{i+1} - X_i) / 2$  are calculated.

9. Organization of calculation cycle /7-8/ changing from  $l$  to 1 up to  $N$ .

10. Setup  $E_l$  with respect to magnetic permeability of medium inside and outside the contour  $l$ .

11. Calculation of  $L$  parameter  $\lambda, / LI$  parameter/.

12. Calculation of  $R3 = 0$ .

13. Verification of the condition  $MAG = 1, T = 1, TA = 1$ , if it is fulfilled, then  $R3 = L \cdot LI / LK$

14. Calculation of  $R = L \cdot D_M / LK, R1 = 0$

15. Verification of the condition  $T = 1$ , if it is fulfilled, then  $RI = R$

16. Verification of the condition  $K = M$ , if it is fulfilled, pass to block 19.

17. Calculation of  $G_{KM}$  – coefficients of the system of algebraic equation.

18. Unconditional jump to block 20.

19. Calculation of  $C_{KM}$  for  $K = M$ .

20. Organization of calculation cycle of the coefficients of the  $K$  - th line of the system of algebraic equations /1.21/ changing  $m$  from 1 to  $N$ .

21. Calculation of  $R2$ .

22. Verification of the condition  $TA = 1$ , if it is fulfilled, pass to block 24.

23. Calculation of  $R2$

24. Calculation of  $CP_K$  – coefficients of the right side of the  $K$  - th line of the system of algebraic equations.

25. Organization of calculation cycle of all  $K$  - lines of the coefficients of the system of algebraic equation and right sides changing  $K$  from 1 to  $N$ .

26. Solving the system of linear algebraic equations by the Gauss method.

27. Calculation of  $RB_i = 0$

28. Calculation of  $RB_i = RB_i + CP_k \cdot D_k$ .

29. Organization of calculation cycle changing  $K$  from 1 to  $N$

30. Printing  $RB_i$

31. Organization of calculation cycle for  $TA / TA = 1$  when only one of the first given field source is active and  $TA = 2$  when both sources are active, changing  $TA$  from 1 to 2 through 1.

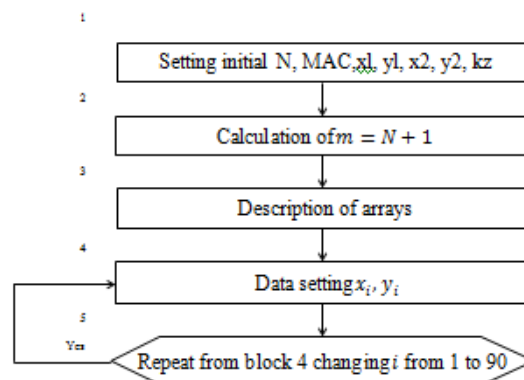
32. Organization of calculation level for  $T/T=1$  when solving the transformed equation  $TA=3$  when solving untransformed equation changing  $T$  from 1 to 3 through 2.
33. Organization of the impus cycle of magnetic permeability changing  $K_1$  from 1 to  $K$  through 1.
34. End.

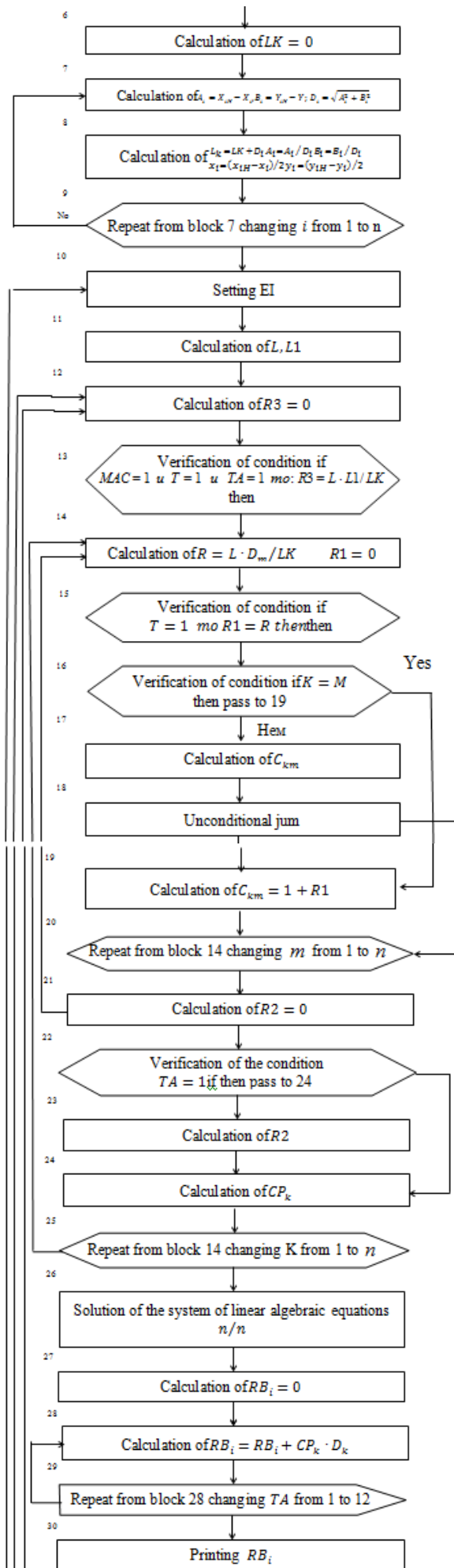
Calculation of flow distribution electrical parameters of the magnetic element of power relay was performed by the given block-diagram of calculation program by MSS (Fig. 2) of the table (1-2).

**Table 1. Errors of calculation of flow distribution in the areas of magnetic element of power relay (fig.2) (number of nodes 38)**

Number of topological areas	Experimental values of magnetic flux $ \Phi_i  10^{-5} B\delta$	Calculation values of magnetic flux $ \Phi_i  10^{-5} B\delta$ by the method of secondary sources	Calculation errors /Exp./-/Calc./ Exp. 100% by the method of secondary sources
1	4,473	4,741	-5,992
2	5,245	5,062	3,489
3	3,768	3,885	-3,105
4	3,617	3,793	-4,866
5	3,820	3,674	3,822
6	4,365	4,215	3,436
7	9,213	9,579	-3,973
8	4,225	4,094	1,101
9	3,586	3,475	3,095
10	3,652	3,850	-5,422
11	3,661	3,784	-3,360
12	4,905	4,665	4,893
13	4,663	4,396	5,115
14	1,504	1,599	-5,964
15	2,438	2,355	3,404
16	2,235	2,369	-5,996
17	2,007	2,073	-3,288
18	3,983	3,475	5,975
19	1,184	1,255	-5,997
20	0,906	0,858	5,298
21	2,122	2,249	-5,985
22	3,755	3,638	3,116
23	0,959	0,905	5,631
24	1,114	1,078	3,232
25	1,817	1,926	-5,999
26	2,135	2,070	3,044
27	2,208	2,077	5,933
28	1,626	1,547	4,859
29	0,251	0,239	4,781
30	0,254	0,240	5,512
31	0,220	0,212	3,536
32	0,218	0,211	3,211
33	0,202	0,194	3,960
34	0,179	0,173	3,352
35	0,262	0,271	-3,435
36	0,280	0,270	3,571
37	0,177	0,184	-3,955
38	0,175	0,169	3,429

### Block-diagram for calculating magnetic elements and devices by MSS





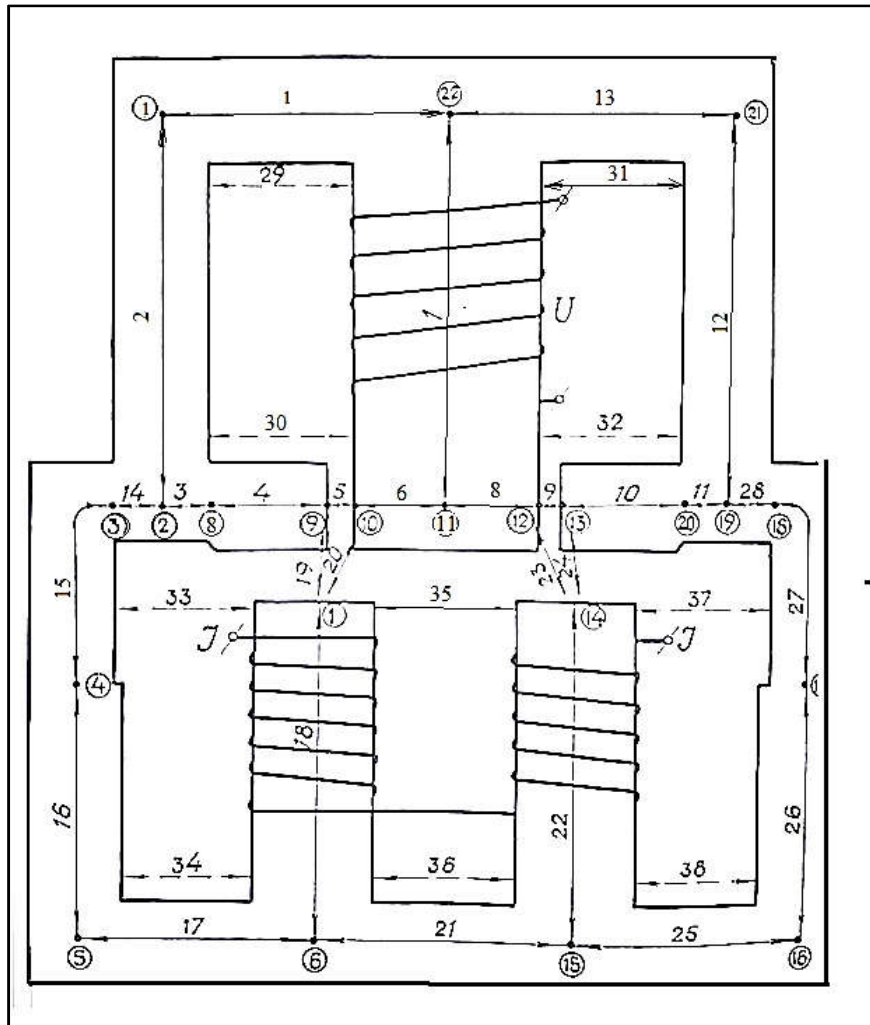
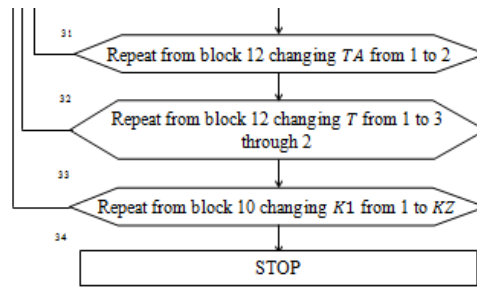


Fig. Magnetic element of power relay divided into areas

## Conclusions

1. Calculation of magnetic elements and devices of automatics of control systems allows completely to analyze the calculation process.
2. Based on the suggested algorithm, the flow distribution and also electric, magnetic parameters of rather complex magnetic elements and devices are determined by elementary areas.
3. The suggested calculation algorithm significantly reduces the labor-intensiveness of the process and also rather virtually reduce the error of calculation with respect to experimental data.
4. The finite element method, as one of various numerical methods allows widespread use of calculating systems for calculating different complex magnetic elements and devices with regard to specificity of these elements with giving rather effective value of calculation.

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